Magnetic Gearing Electromagnetic Concepts

C. G. C. Neves, A. F. F. Filho Curso de Engenharia de Energia/Universidade Federal do Pampa – Bagé – Brazil

cgcneves@ieee.org

Dep. Eng. Elétrica/DELET - Universidade Federal do Rio Grande do Sul - Porto Alegre, Brazil <u>alv.flores@ufrgs.br</u>

Abstract— In this paper the magnetic geared machines electromagnetic concepts are introduced. The magnetic gear modulation mechanism which becomes possible rotors of different polarities produce torque is explained, through the use analytical expressions and plots obtained by finite elements method. This pedagogical approach is a way to teach the subject in a more understandable way and, at the same time, it reinforces the application of electromagnetic concepts. Further, a magnetic geared generator is simulated and no-load EMFs and torques are analyzed.

Index Terms— magnetic gear; permanent magnet generator; wind power; eletromagnetic concepts; cogging torque

I. INTRODUCTION

Many papers about magnetic gears and derived geared machines as magnetic geared generators were published and market applications started to appear [4] [5] [6] [7] [8]. However, in spite of the excellent level of the papers published on the subject, they are of difficult understanding for the average-level student. This is quite visible on the mechanism of modulation responsible for creation the flux densities harmonics which interact to generate the torque in the geared machines.

II. ELECTRICAL MACHINES TORQUE PRODUCTION

The operation of a rotating electric machine can also be regarded as an interaction between two magnetic fields [1]. The machine action is due to the result of these two fields trying to line up, so that the central line of a north pole on one part is directly opposite the central line of a south pole on the other part. Torque is produced by the interaction of the two magnetic fields. Figure 1(a) shows the simplified construction of a two-pole machine, in which the north and south poles of the rotor are attracted by south and north poles (and repelled by north and south poles, respectively) of the stator, resulting in a torque in the counter-clockwise direction. The magnitude of the torque is proportional to the product of the field strengths of the two fields. The angle δ between the axes of the two fields is known as torque or power angle. For sinusoidal variation of flux in the air gap, i.e. flux density varying sinusoidally with distance around the gap periphery, the torque is also proportional to δ . However, if the number of poles on rotor and stator are different, as shown in Figure 1(b) (machine with 4 poles in the rotor and 2 poles in the stator) the net torque is always zero with respect to the axis of the stator field. On the N1N2 axis, the pole N1 is repelled by pole N and attracted by pole S

producing a counter-clockwise torque on the rotor. The pole N2 is repelled by N pole and attracted by S pole producing an equal clockwise torque. Hence, the net torque is zero. A similar situation takes place on S1S2 axis. Therefore, no net electromagnetic torque is produced. If the machine has 2 poles in the rotor and 4 poles in the stator, as in Fig. 1(c), the same occurs. Whenever the number of poles on rotor and stator are different, net torque is always zero.

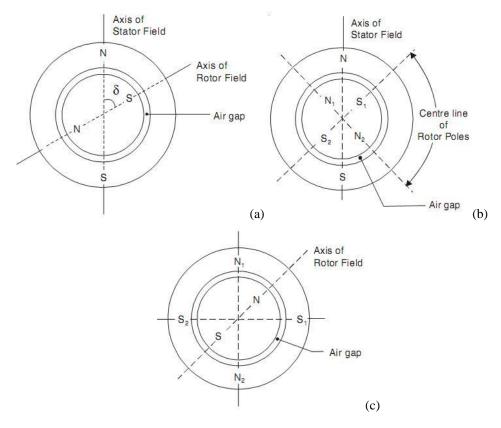


Fig. 1. Interaction of stator and rotor magnetic fields: (a) 2 poles in the rotor and 2 poles in the stator. (b) 4 poles in the rotor and 2 poles in the stator. (c) 2 poles in the rotor and 4 poles in the stator.

The conclusion is that all rotating machines (generators and motors) must have the same number of poles on the stator and rotor for steady unidirectional torque.

III. MODULATION PHENOMENA OPERATING PRINCIPLE

The operating principle of the flux-modulation can be explained with the help of Fig. 2, which is a simplified picture of the outer rotor composed with PMs polarized north and poles interacting with the static modulators (ferromagnetic pole pieces) [2]. The left hand side of each picture represents the PMs on the outer rotor. The middle black pieces represent the ferromagnetic flux modulator and the right-hand side represents the flux content produced by modulation that the inner rotor would interact with. This means the flux at left hand side is created by the outer PMs while the flux at right hand side is created by modulation. In Fig. 2(a) the modulation pieces will only produce flux from north poles, because they face only north poles, resulting in a north pole. In Fig 2(b) the modulation pieces will only produce flux from south poles, because the modulation pieces are moved just by one space and the outer rotor faces a south pole. In Fig. 2(c) the modulation pieces are spaced differently, so the middle

pairs of north and south poles will oppose each other, and only the end ones will produce flux creating one north and one south at the ends that would interact with the inner rotor. The relative position between the outer rotor PMs and modulators controls how the two rotors interact with each other.

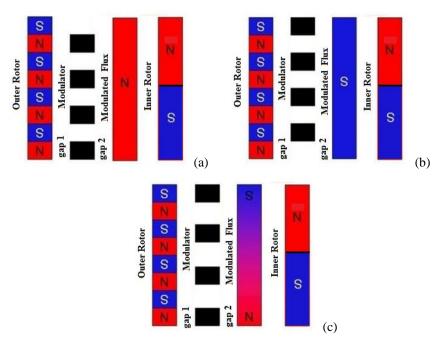


Fig. 2. Power Interaction of outer rotor PMs and modulators: (a) Only a north pole. (b) Only a south pole. (c) A north-south pair which will interact with the inner rotor north-south pair.

IV. MAGNETIC GEARED MACHINES CONSTRUCTION

A Magnetic Gear consists of following components as shown in Fig. 3(a):

- 1. Outer permanent-magnet array (OPMA) with a pole-number (p_l).
- 2. Inner permanent-magnet array (IPMA) with a number of pole-pairs (p_h).
- 3. Intermediate annular pole-pieces (IAP) composed of Ferro-magnetic modulators (np).

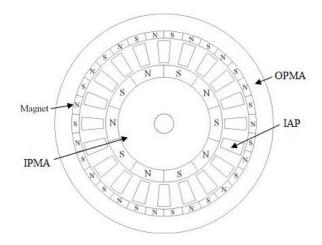


Fig. 3. Construction of a Magnetic Gear.

And the Magnetic Geared Generator consists of the same components of the magnetic gear with the addition of a stator juxtaposed to OPMA, comprising a slotted lamination pack (SLP) as shown in Fig. 4:

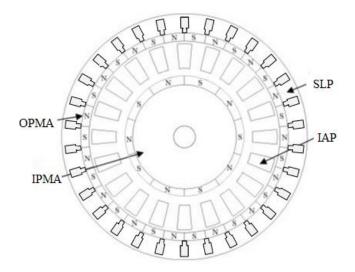


Fig. 4. Construction of Magnetic Geared Generator.

V. CO-AXIAL MAGNETIC GEAR WORKING PRINCIPLE

If the reasoning applied on the section II is used in a cylindrical rotating structure, one can notice that the situation shown at Fig. 2(c) can be reproduced. Figure 3(a) shows a cylindrical magnetic structure composed of two rotors separated by an air-gap, the outer rotor and the inner rotor with different number of poles. Since the numbers of poles are different, it is not possible to produce torque. However, if the ferromagnetic pole pieces (modulators) are putted between the outer rotor and the inner rotor, it will alternately "short-circuit" N-S pole-pairs of outer rotor or produce a pole (north or a south) similar to what occurred at Fig. 2(c). The result of this is a modulated 10 poles flux in the inner air-gap, represented by the red North's and blue South's, shown on Fig. 4(b) since the inner rotor has 10 poles the net torque will be different than zero and constant as explained in the first section of this paper.

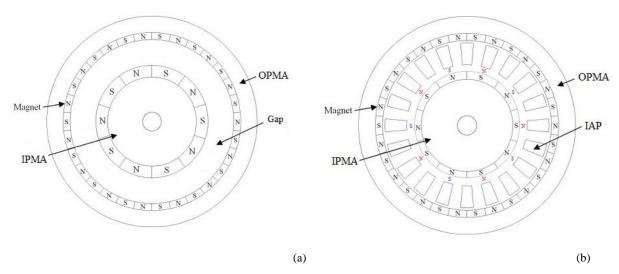


Fig. 4. (a) Numbers of poles are different, it is not possible to produce torque. (b) Modulators action: Modulated 10 poles flux in the inner air-gap, represented by the red North's and blue South's.

VI. MAGNETIC GEARED GENERATOR WORKING PRINCIPLE

The principle of operation of the magnetic geared generator can be understood by considering the operation of the different sub-components within it; that is, the magnetic gear and the electrical machine. Fig. 5(a) shows the elements which contribute to its magnetic gearing action: the IPMA, IAP and OPMA. Fig. 5(b) shows the elements which contribute to its working as an electrical machine: the SLP with the copper windings and the IAP.

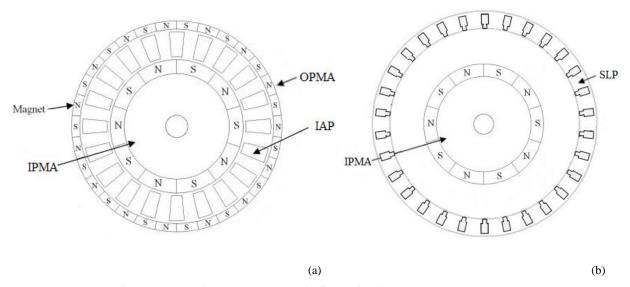


Fig. 5. (a) Magnetic gear components. (b) Convectional PM generator components.

The high-speed inner rotor or IPMA contributes to the operation of both the electrical machine and the magnet gear. This rotor interconnects the magnetic gear with the electrical machine and so the mechanical torque is not applied to the IPMA, but on IAP which is the input shaft. The mechanical torque applied to the IPMA by the magnetic gear is smaller than that applied to the IAP due to the gearing effect. Consequently, the electromagnetic torque produced by the stator could be only a fraction of the total input torque. Thus the required magneto-motive force (MMF) in the machine, and hence the copper losses, are significantly smaller than in a conventional direct-drive generator, which would have to react the full torque on the input shaft [3]. Consequently the geared machine uses less active material than would be used by a combination of a magnetic gear and separate electrical machine.

VII. MAGNETIC GEAR AIR GAP FLUX DISTRIBUTION

Flux harmonics are essential for the operation of a geared machine then they will be discussed in detail in this section. The flux distribution in the outer air gap as function of time will be mathematically expressed and studied when the OPMA is removed and only the IPMA and IAP remain. The respective rotation velocities of the IPMA and IAP are given as ω_h and ω_p for the magnetic gear shown at Fig. 6.

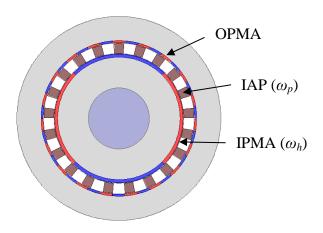


Fig. 6. Simplified view of the coaxial magnetic gear.

The magneto motive force (MMF) of the IPMA and the permeance of the IAP are mathematically expressed by (1) and (2), respectively

$$F_{ri}(\theta) = \sum_{h=1}^{\infty} \frac{A}{(2h-1)} \sin\{(2h-1)p_h(\theta-\omega_h t)\}$$
 (1)

$$P_r(\theta) = P_0 + \sum_{\ell=1}^{\infty} \frac{B}{(2\ell - 1)} \sin\{(2\ell - 1)n_p(\theta - \omega_p t)\}$$
 (2)

where A is the permanent magnet MMF amplitude, B is the modulators permeance amplitude and P_0 is the permeance of vacuum.

The product of (1) and (2) yields the outer air gap flux distribution:

$$\phi_{o}(\theta) = \sum_{h=1}^{\infty} \frac{AP_{0}}{(2h-1)} \sin(C_{1}\theta - \omega_{h}t) + \frac{1}{2}C_{0} \sum_{h=1}^{\infty} \sum_{p=1}^{\infty} \circ \begin{cases} \cos(C_{2} - C_{1}) \left(\theta - \frac{C_{2}\omega_{p} - C_{1}\omega_{h}}{C_{2} - C_{1}}t\right) \\ -\cos(C_{2} + C_{1}) \left(\theta - \frac{C_{2}\omega_{p} + C_{1}\omega_{h}}{C_{2} + C_{1}}t\right) \end{cases}$$
(3)

where
$$C_0 = \frac{AB}{(2h-1(2p-1))}$$
, $C_1 = (2h-1)p_h$ and $C_2 = (2p-1)n_p$.

From (3), it can be seen that flux distribution consists of three main components: one fundamental component and two harmonic components $C_2 \mp C_I$. The fundamental component (FC) depends only of $F_{ri}(\theta)$, given by Eq. (1) which depends of spatial harmonic number C_I and of rotation speed ω_h , that are exactly the same speed of $F_{ri}(\theta)$. The other two harmonic components (called H1 and H2) exist due to the permeance of the IAP pole-pieces which modulates $F_{ri}(\theta)$. Also, from (3) it can be seen that the angular velocities of H1 and H2 are different from that of the MMF. This means that when the IPMA is rotated, the fundamental (FC) follows the IPMA, but H1 and H2 do not follow it. This is the main point of the operating principle of a magnetic gear. Either H1 or H2 can be chosen to create a magnetic gear or the magnet gearing effect, but usually H1 is chosen due to its higher flux density. Figure 7(a) shows the computed flux distribution in the outer air gap at t= 0 sec using FEM and its harmonic analysis is shown in Fig. 7(b).

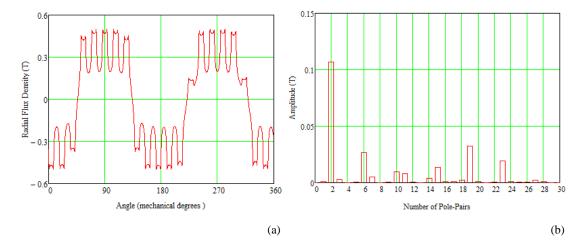


Fig. 7. (a) Radial field density at outer air-gap due to the inner magnets only, with modulating rotor. (b)Radial field density harmonic content at outer air-gap due to the inner magnets only, with modulating rotor.

The dominant components are 2^{nd} , 19^{th} , and 23^{th} , which each corresponds to FC, H1 and H2, respectively, since $p_h = 2$ and np = 21, when h = p = 1 in Eq. 3. Since H1 has much higher flux density than H2, p_l must be equal to $19 (n_p - p_h = 19)$. Taking this as general rule the following generic poles relation is adopted:

$$p_h + p_l = n_p \tag{4}$$

VIII. MAGNETIC GEARED MACHINES GEAR RELATIONS

If each term of (4) is multiplied by its respective speed, it gets the following expression:

$$p_h \omega_h + p_l \omega_l = n_p \omega_p \tag{5}$$

Where, ω_h is speed of inner rotor, ω_l is speed of the outer rotor and ω_p is the speed of modulator. When one of the three parts of the gear is stationary, there will be a constant relation or *gear ratio* (G_r) between the speeds of other two parts. For example when the modulators are stationary $(\omega_p = 0)$ the machine works as mechanical gear, G_r is equal to 9.5 since $p_l = 19$ and $p_h = 2$:

$$G_r = \frac{\omega_h}{\omega_l} = -\frac{p_l}{p_h} \tag{6}$$

Therefore, the relation between the speeds is inversely proportional to the ratio between the numbers of pole-pairs, and rotors run in opposite directions. If it is considered there are no losses, the power in inner rotor is equal to power in outer rotor, then:

$$\frac{T_h}{T_l} = -\frac{\omega_l}{\omega_h} = -\frac{p_h}{p_l} = \frac{1}{G_r} \tag{7}$$

Therefore, the ratio between the torques is directly proportional to ratio between the numbers of pole-pairs. Alternatively, when the OPMA is held stationary ($\omega_l = 0$), and the IAP is driven by a turbine the machine works as magnetic geared generator, the gear ratio becomes:

$$G_r = \frac{\omega_h}{\omega_p} = \frac{np}{p_h} \tag{8}$$

Therefore, the relation between the speeds is inversely proportional to the ratio between the numbers of modulators and numbers of pole-pairs of IPMA, and both rotors run in the same direction. If it is considered there are no losses, the power in inner rotor is equal to power in the modulators, then:

$$\frac{T_p}{T_h} = \frac{\omega_h}{\omega_p} = \frac{np}{p_h} = G_r \tag{9}$$

As both shafts are now rotating in the same direction, a higher gear ratio of 10.5 is obtained since np = 21 and $p_h = 2$.

IX. MAGNETIC GEARED GENERATOR SIMULATION AND ANALYSIS

If slots are stamped in the stator of the magnetic gear and they are wounded in a certain way the magnetic gear became a magnetic geared generator as shown on Fig. 8 (a). In the geared generator model considered p_h , p_l and np are equal to magnetic gear analyzed in the last section, then the 19^{th} (H1) is used to create the gearing effect. However, a harmonic flux component of air gap must couple with the stator winding to electromotive-motive-forces (EMFs) be induced in the winding phases. This is done by ensuring that the number of generator winding pole pairs is the same as the IPMA array pole pair, then in the operation as this magnetic geared generator the 2^{nd} flux component couples with the generator coils to produce power as shown in Fig. 8 (a). In order to increase the 2^{th} harmonic component and increase the EMFs, in this model, there are 6 generator coils and they are wound as show in Fig. 8 (b).

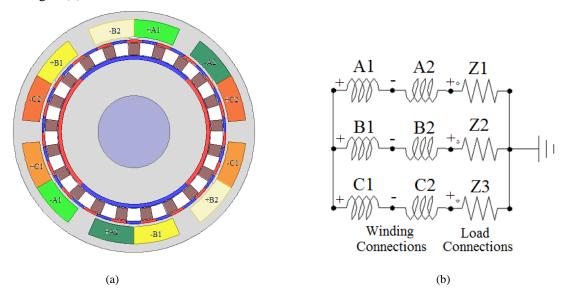


Fig. 8. (a) Simplified view of the Magnetic Geared Generator. (b) Magnetic Geared Generator winding connections.

A. Stator Winding Flux Analysis

In order to calculate only the flux generated by stator winding the OPMA and IPMA were removed, and the flux distribution of the air gap inside of the stator when only the generator coils were excited was determined by FEM static analysis. Figure 9(a) shows the resulting waveform and Fig. 9(b) the corresponding harmonic spectrum. From it can be seen that the 2thcomponent is the dominant

harmonic. Therefore it can be confirmed that the generator coils will be able to be excited by the IPMA permanent magnets and EMFs will induced in the stator winding phases.

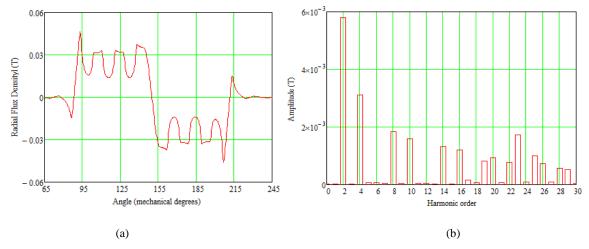


Fig. 9. (a) Flux density of Stator. (a) Waveform (b) Harmonic Spectrum

B. Electromagnetic Torque Modelling

Electromagnetic torque is produced in a magnetic geared machine when the energy stored in magnetic fields varies with rotation of the components. Torque on a component can be expressed as:

$$T = \frac{\partial W'}{\partial \theta_{\alpha}} \tag{10}$$

where is W the co-energy of the system and is the θ_c angular position of the component on which the torque is calculated. In this analysis it is assumed that all the co-energy is stored in the air-gaps and the magnets. In these regions, the co-energy can be calculated as

$$W' = \frac{1}{2\mu_0} \int_V B^2 dV \tag{11}$$

where B is the air-gap flux density and assuming that the permeability of the magnets is approximately equal to that of air (free space).

The movement is taking into account by Movement Band Technique [9] and the Coulomb's virtual work method [11] was used for torque calculation in the moving band:

$$T = \frac{L}{\mu_0} \sum_{e=1}^{N_{mb}} \int_{\Omega_e} \left(-\mathbf{B}^T \mathbf{G}^{-1} \frac{\partial \mathbf{G}}{\partial \theta} + \frac{1}{2} \frac{B^2}{|\mathbf{G}|} \frac{\partial |\mathbf{G}|}{\partial \theta} \right) d\Omega$$
 (12)

where $\mathbf{B} = [B_x, B_y]$, $B = \|\mathbf{B}\|$, \mathbf{G} the Jacobian matrix of the global nodal coordinates with respect to local element coordinates, and $[\mathbf{G}]$ the determinant of \mathbf{G} .

The static torque on inner gap is obtained holding the OPMA and IAP still and incrementally rotating the IPMA step by step (rotating at 1 degree/sec) through the movement band shown in Fig. 10(a). The magnetic torque on outer gap is obtained in the same way but the movement band assumes the position shown in Fig. 10(b). The torque on IAP is equal to difference between the torque calculated at outer gap and inner gap [12, 13].

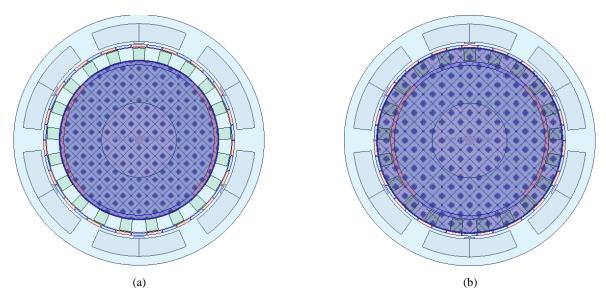


Fig. 10. (a) Band to calculate torque at inner gap (rotates at 1 degree/sec). (b) Band to calculate torque at outer gap (rotates at 1 degree/sec).

Figure 11(a) shows the torque on the three regions. It can be found that the torque-angle curves vary sinusoidally, in which the maximum torque values denote the pull-out torques. On the IAP and inner gap, the pull-out torques is 475.88 Nm and 44.88 Nm, respectively. Their ratio is 10.6, which has a good agreement with the ratio n_p/p_h that is equal to10.5.

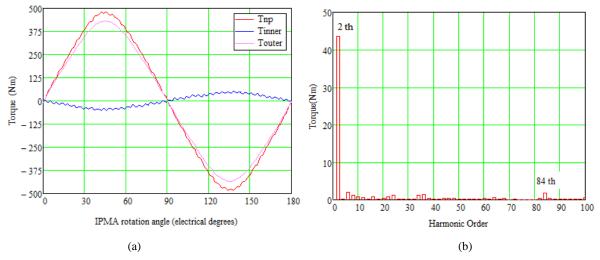


Fig. 11. (a) Static torque waveforms. (b) Static torque harmonic spectrum at inner air-gap (output torque).

C Cogging Torque Analysis

The cogging torque manifests itself as torque ripple harmonics on the inner air gap (output rotor) due to interactions between the IPMA and the IAP, when windings currents are null. For the inner rotor the fundamental cogging torque harmonics equal to 84 pole-pairs, this means, the smallest common multiple between the number poles of inner rotor, $2p_{in} = 4$, and the number of ferromagnetic pole-pieces, $n_p = 21$. The cogging factor is equal to $2p_h ns/Nc$, which gives 1. The harmonic

spectrum shown at Fig. 11(b) complements this reasoning since the value of the 84th harmonic is only 2% of the fundamental (2th). Hence, the cogging torque is inherently small.

D No-load EMF

In order to simulate the movement of both rotors in relation to the winding and modulators and obtain no-load EMF two movement bands are applied as shown on Fig. 12. The external band, placed in the IAP (input rotor), rotates at 150 rpm and the internal band, placed in the inner air-gap (out rotor) at 1575 rpm, since the speeds relation is 10.5. As already explained at section VIII both bands must run in the same direction.

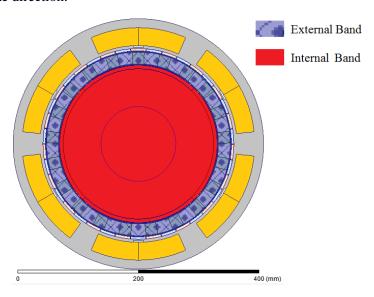


Fig. 12 Bands to calculate EMFs (inner band rotates at 1575 rpm and outer band rotates at 150 rpm). Both bands must run in the same direction

Figure 13 shows the 3-phase EMFs induced when the OPMA is held stationary ($\omega_l = 0$), the stator winding is open, and the IAP rotates at 150 rpm simulating the speed transmitted by the shaft of a wind turbine. The waveform is non-sinusoidal due to the large odd harmonic (3^{rd} , 5^{th} and 7^{th}) as can be seen at harmonic spectrum shown at Fig. 13(b). One can infer that this harmonics are due to the generators' winding configuration since the harmonic spectrum of waveform produced only by the stator winding, shown at Fig. 8(b) presents odd harmonics too.

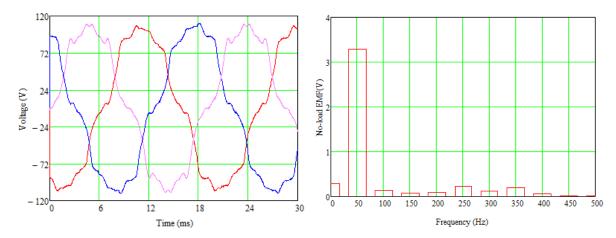


Fig. 13 No-load EMFs (a) Voltage waveforms. (b) Frequency Harmonic Spectrum.

X. CONCLUSION

In this paper the magnetic geared machines electromagnetic concepts are introduced. The magnetic gear modulation mechanism which makes possible rotors of different polarities produce unidirectional torque was explained in a simple way. The magnetic geared generator working principle was explained. The poles and speed relations of the magnetic geared machines are obtained and explained through the use analytical expressions and plots obtained by finite elements method. The gear relations for magnetic gears and magnetic geared generators are demonstrated mathematically. The geared generator simulated presents low cogging torque but no-load EMFs generated are deformed by odd harmonics generated by the stator winding.

REFERENCES

- [1] Umans, S. Fitzgerald and Kingsley's Electric Machinery, McGraw-Hill, 2013.
- [2] Brönn, L. Design and Performance Evaluation of a Magnetically Geared Axial-Flux Permanent Magnet Generator. Thesis: MScEng (EE),2012.
- [3] Chau, K. T., "A Novel Coaxial Magnetic Gear Using Bulk HTS for Industrial Applications," *IEEE Trans.* Appl. Supercond., vol. 20, no. 3, pp. 981–984, Jun. 2010.
- [4] Jian, L., K.T. Chau, "Analytical calculation of magnetic field distribution in coaxial magnetic gears", Progress In Electromagnetics Research, Vol. 92, pp. 1-16, 2009.
- [5] Atallah, K. and D. Howe, "A novel high-performance magnetic gear," IEEE Transactions on Magnetics, Vol. 37, No. 4, pp. 2844-2846 2001
- [6] Lian, L; Chau, K.-T; Gong, Y; Jiang, JZ; Yu, C; Li, W, "Comparison of coaxial magnetic gears with different topologies", IEEE Transactions On Magnetics, , Vol. 45, n. 10, p. 4526-4529, 2009
- [7] Atallah, K., S. D. Calverley, and D. Howe, "Design, analysis and realization of a high-performance magnetic gear," IEE Proc. Electr. Power Appl., Vol. 151, No. 2, pp. 135-143, 2004.
- [8] X. Li, K.-T. Chau, M. Chengand, W. Hua, "Comparison of magnet-geared permanent-magnet machines," Progress In Electromagnetics Research, Vol. 133, No. 3, pp. 177-198,2013
- [9] B. Davat, Z. Ren and M. Lajoie-Mazenc, "The movement in field modeling," IEEE Trans. Magn., vol. 21, no. 6, pp. 2296–2298, November 1985.
- [10] B. Silwal, P. Rasilo, L. Perkkio, A. Hannukainen, T. Eirola and A. Arkkio, "Evaluation and Comparison of different numerical computation methods for the electromagnetic torque in electrical machines," in Proc. Int. Conf. Electrical Machines and Systems, Oct 2013, pp.837–842.
- [11] J. Coulomb and G. Meunier, "Finite element implementation of virtual work principle for magnetic or electric force and torque computation," IEEE Trans. Magn., vol. 20, no. 5, pp.1894–1896, 1984.